



신호 및 시스템

III. FIR 필터



- 학습목표
 - FIR 필터의 개념 및 특성을 이해한다.
 - FIR 필터의 구현 방법을 학습한다.
 - FIR 필터의 주파수 응답을 구한다.

Discrete-time System



- Operate on $x[n]$ to get $y[n]$
- A general class of systems
 - **ANALYZE** the system
 - Tools: time-domain & frequency-domain
 - **SYNTHESIZE** the system
- Example
 - Running average
 - **RULE**: “the output at time n is the average of three consecutive input values”

General FIR Filter



- Filter coefficients $\{b_k\}$

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$

- Filter order is M
- Filter length is $L = M+1$
- Number of filter coefficients is L

- Example

- 3-pt Average System $y[n] = \frac{1}{3}(x[n] + x[n+1] + x[n+2])$

- $b_k = \{3, -1, 2, 1\}$ $y[n] = \sum_{k=0}^3 b_k x[n-k]$
 $= 3x[n] - x[n-1] + 2x[n-2] + x[n-3]$

Example: filtered stock signal



- 50-pt Averager

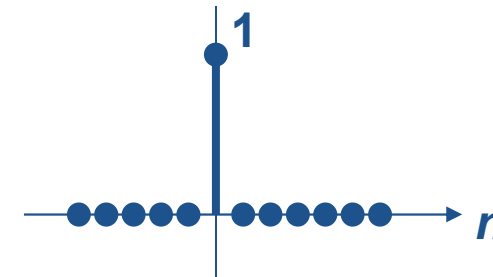


Special Input Signals (1/2)



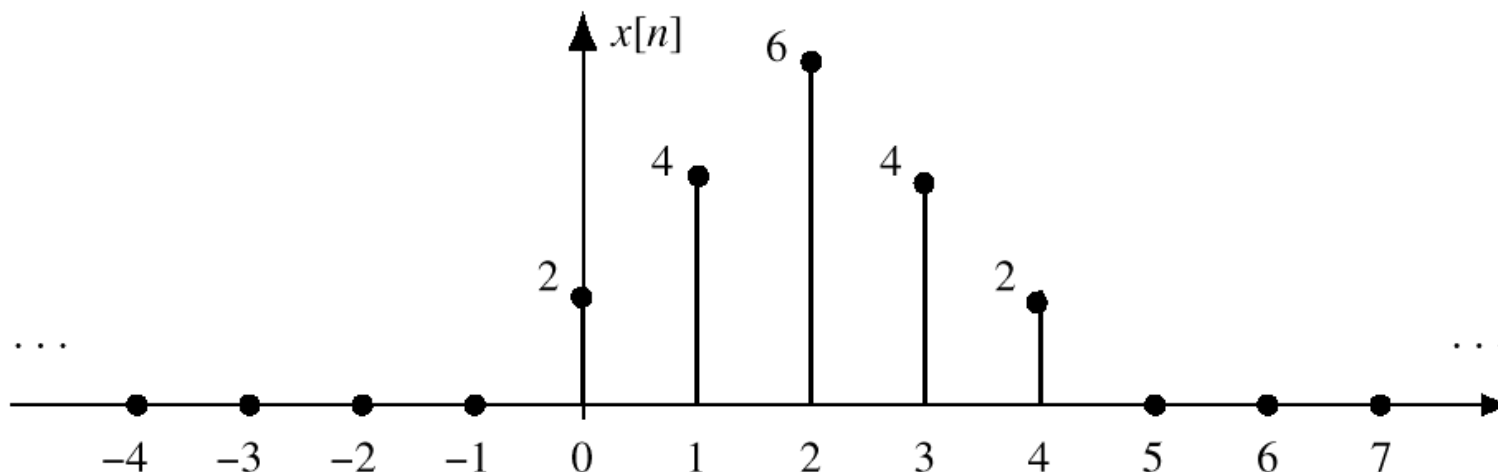
- Unit impulse
 - The mathematical notation is that of the Kronecker delta function.

$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$



$$x[n] = \sum_k x[k] \delta[n - k] = \dots + x[-1] \delta[n + 1] + x[0] \delta[n] + x[1] \delta[n - 1] + \dots$$

$$x[n] = 2\delta[n] + 4\delta[n - 1] + 6\delta[n - 2] + 4\delta[n - 3] + 2\delta[n - 4]$$



Special Input Signals (2/2)



- Unit Impulse response
 - When the input to the FIR filter is a unit impulse sequence, $x[n]=\delta[n]$, the output is, by definition, the **unit impulse response**, which is denoted by $h[n]$.

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$



$$h[n] = \sum_{k=0}^M b_k \delta[n-k]$$

⇒ Since $h[n]=0$ for $n<0$, the system is called a **finite impulse response** (FIR).

$$y[n] = \sum_{k=0}^M h[k] x[n-k] \quad \Rightarrow \quad \text{Convolution}$$

Convolution



- Convolution and FIR Filters

- Notation:

$$y[n] = h[n] * x[n]$$

Finite limits

- FIR case:

$$y[n] = \sum_{k=0}^M h[k]x[n-k]$$

Same as b_k

Finite limits



Example: Convolution



$$h[n] = \delta[n] - \delta[n - 1] + 2\delta[n - 2] - \delta[n - 3] + \delta[n - 4]$$

$$x[n] = u[n]$$


$$y[n] = \sum_{k=0}^M h[k]x[n - k]$$

n	-1	0	1	2	3	4	5	6	7
$x[n]$	0	1	1	1	1	1	1	1	...
$h[n]$	0	1	-1	2	-1	1	0	0	0
$h[0]x[n]$	0	1	1	1	1	1	1	1	1
$h[1]x[n - 1]$	0	0	-1	-1	-1	-1	-1	-1	-1
$h[2]x[n - 2]$	0	0	0	2	2	2	2	2	2
$h[3]x[n - 3]$	0	0	0	0	-1	-1	-1	-1	-1
$h[4]x[n - 4]$	0	0	0	0	0	1	1	1	1
$y[n]$	0	1	0	2	1	2	2	2	...

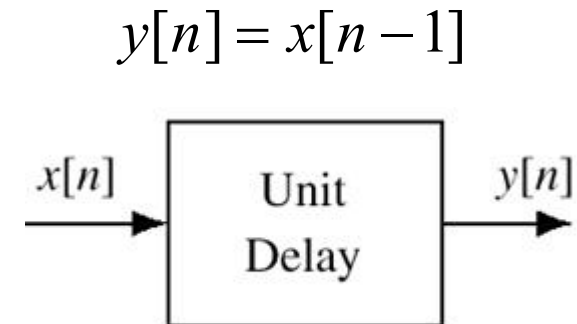
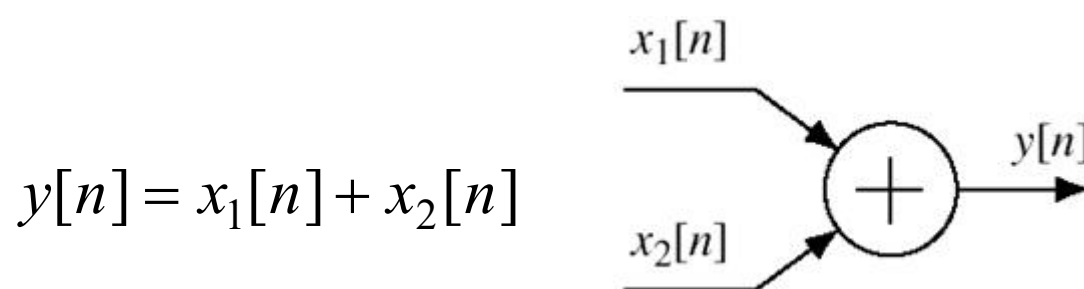
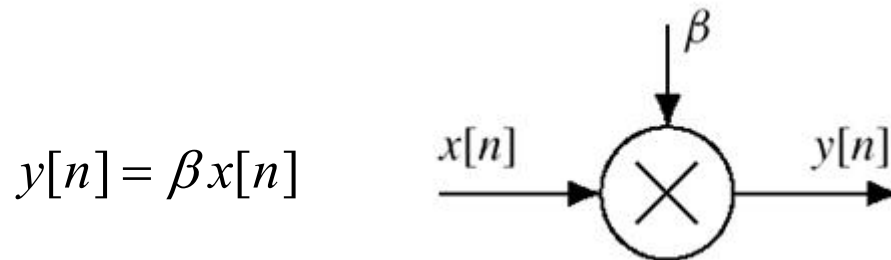
Implementation of FIR filters



- Recall the general definition of an FIR filter

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$


- The basic building-block systems we need are the multiplier, the adder, and the unit-delay operator.



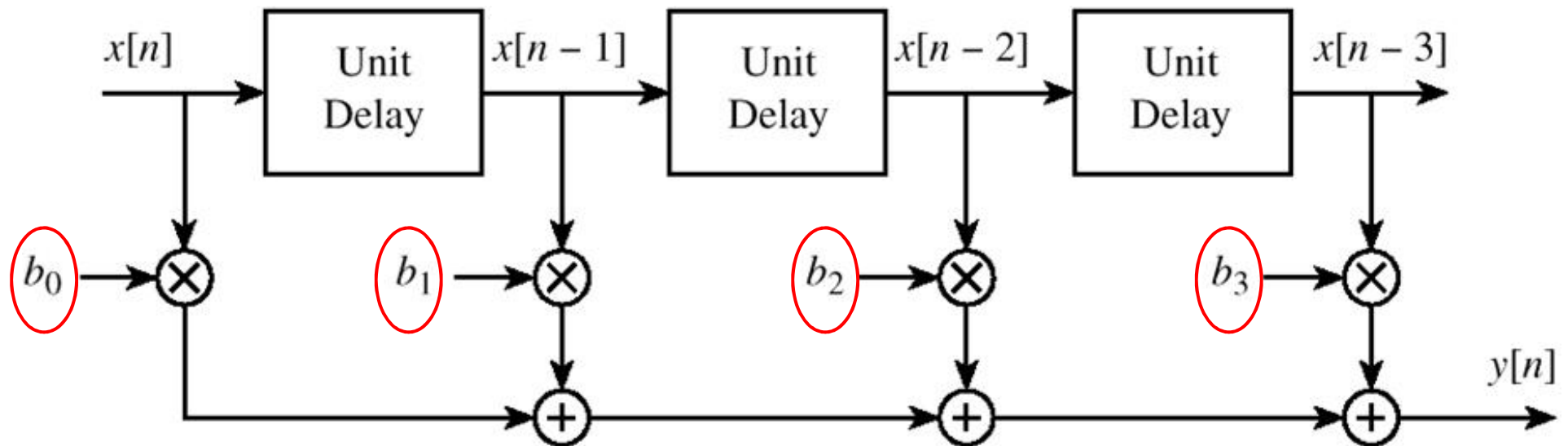
FIR Structure



- Direct Form

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

Signal flow graph

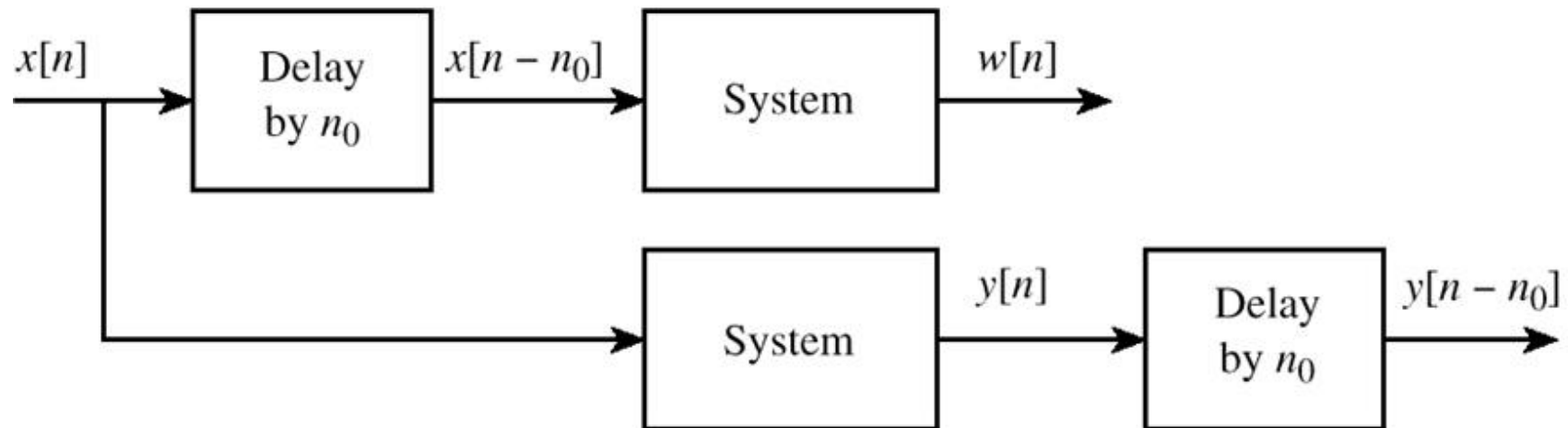


Block-diagram structure for the M th order FIR filter

Linear Time-Invariant (LTI) Systems (1/2)



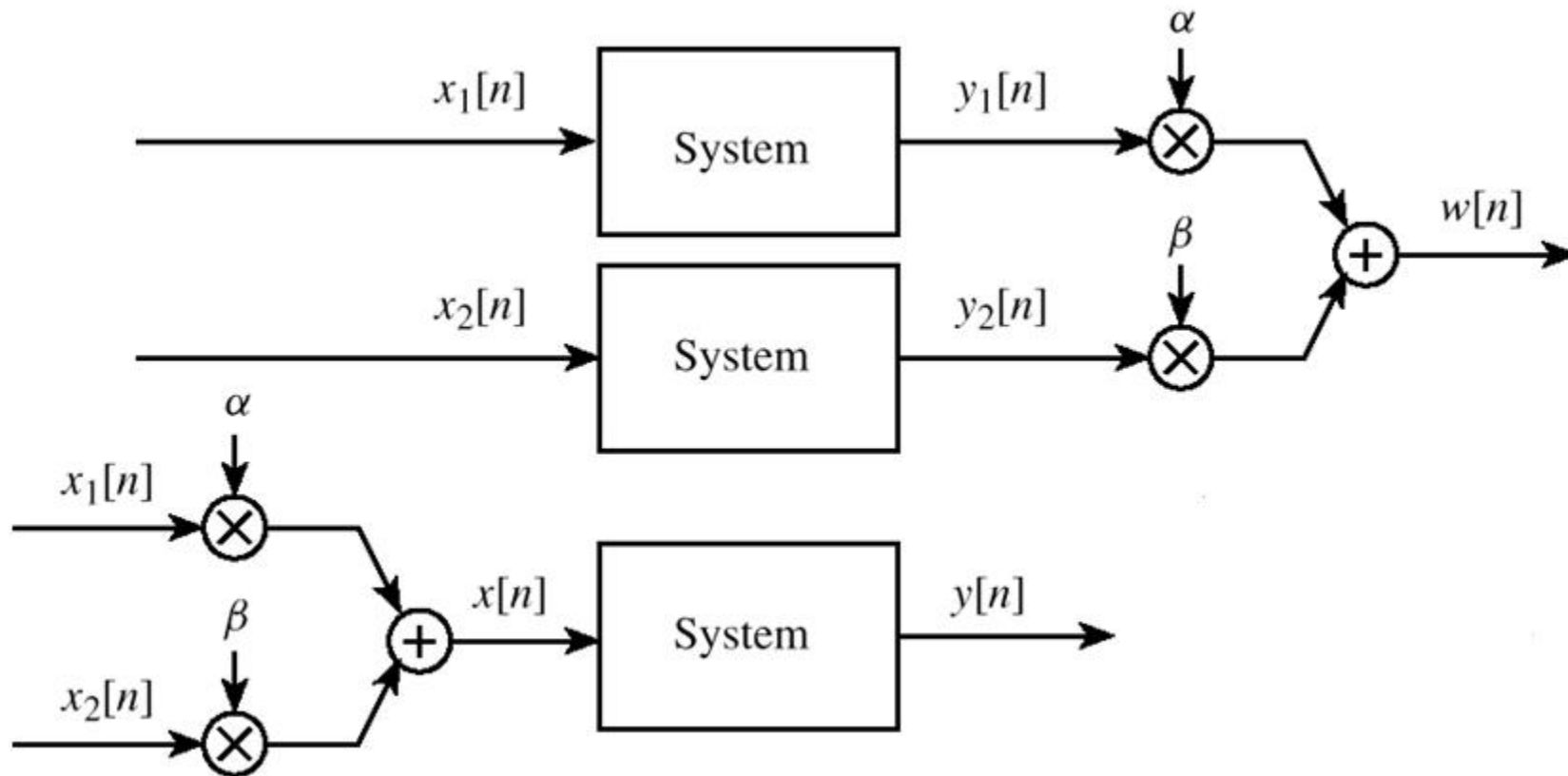
- Time Invariance
 - “Time-Shifting the input will cause the same time-shift in the output”



Linear Time-Invariant (LTI) Systems (2/2)



- Linearity = Two Properties
 - SCALING: “Doubling $x[n]$ will double $y[n]$ ”
 - SUPERPOSITION: “Adding two inputs gives an output that is the sum of the individual outputs”



LTI systems



- Completely characterized by

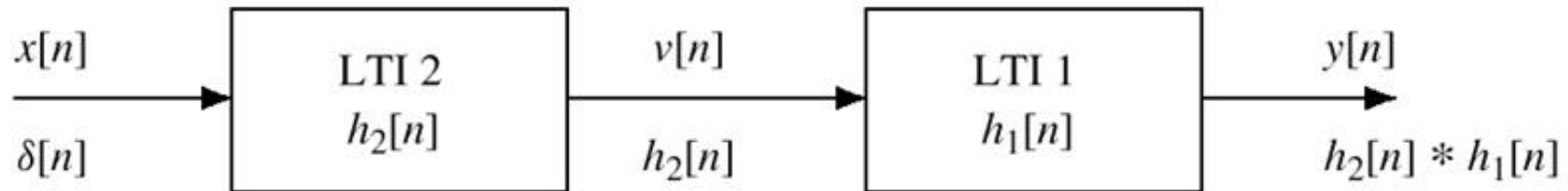
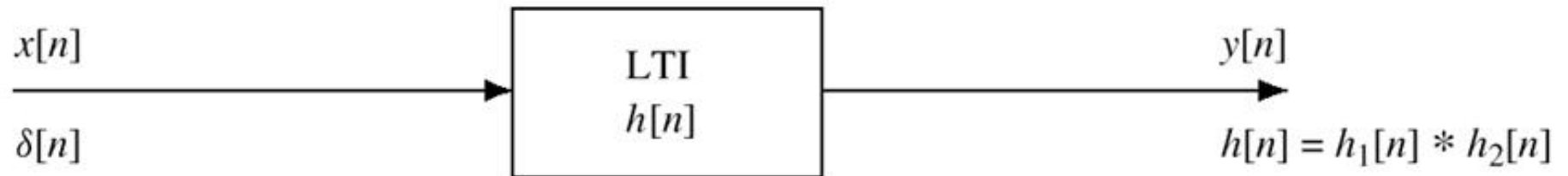
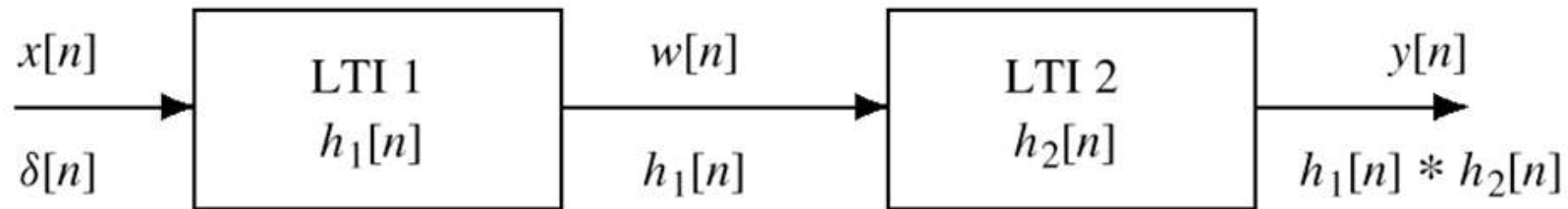
- Impulse response $h[n]$
- Convolution $y[n] = x[n]*h[n]$
- FIR Example: $h[n]$ is same as b_k

- Properties of LTI systems

- Commutative $x[n]*h[n] = h[n]*x[n]$

- Associative $(x_1[n]*x_2[n])*x_3[n] = x_1[n]*(x_2[n]*x_3[n])$

Cascaded LTI systems



Frequency Response of an FIR System



- Sinusoidal Response of FIR systems

$$x[n] = Ae^{j\varphi} e^{j\hat{\omega}n} \quad -\infty < n < \infty$$

$$\begin{aligned} y[n] &= \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M b_k Ae^{j\varphi} e^{j\hat{\omega}(n-k)} \\ &= \left(\sum_{k=0}^M b_k e^{j\hat{\omega}(-k)} \right) Ae^{j\varphi} e^{j\hat{\omega}n} = H(e^{j\hat{\omega}}) x[n] \end{aligned}$$

- Frequency-response

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k} = \sum_{k=0}^M h[k] e^{-j\hat{\omega}k}$$

Example 1 (1/2)



- Show the frequency response of an FIR filter with coefficients

$$\{b_k\} = \{1, 2, 1\}$$

$$\begin{aligned} H(e^{j\hat{\omega}}) &= 1 + 2e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} \\ &= e^{-j\hat{\omega}} (e^{j\hat{\omega}} + 2 + e^{-j\hat{\omega}}) \\ &= e^{-j\hat{\omega}} (2 + 2\cos \hat{\omega}) \end{aligned}$$

Since $(2 + 2\cos \hat{\omega}) \geq 0$

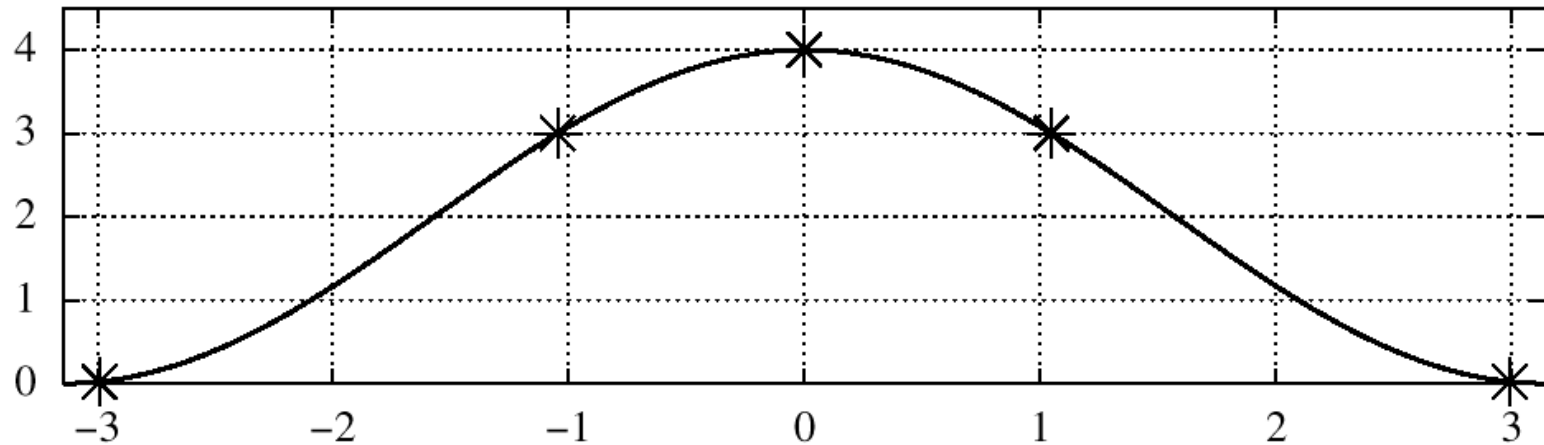
Magnitude is $|H(e^{j\hat{\omega}})| = (2 + 2\cos \hat{\omega})$ and Phase is $\angle H(e^{j\hat{\omega}}) = -\hat{\omega}$

Example 1 (2/2)



$$|H(e^{j\hat{\omega}})| = (2 + 2 \cos \hat{\omega})$$

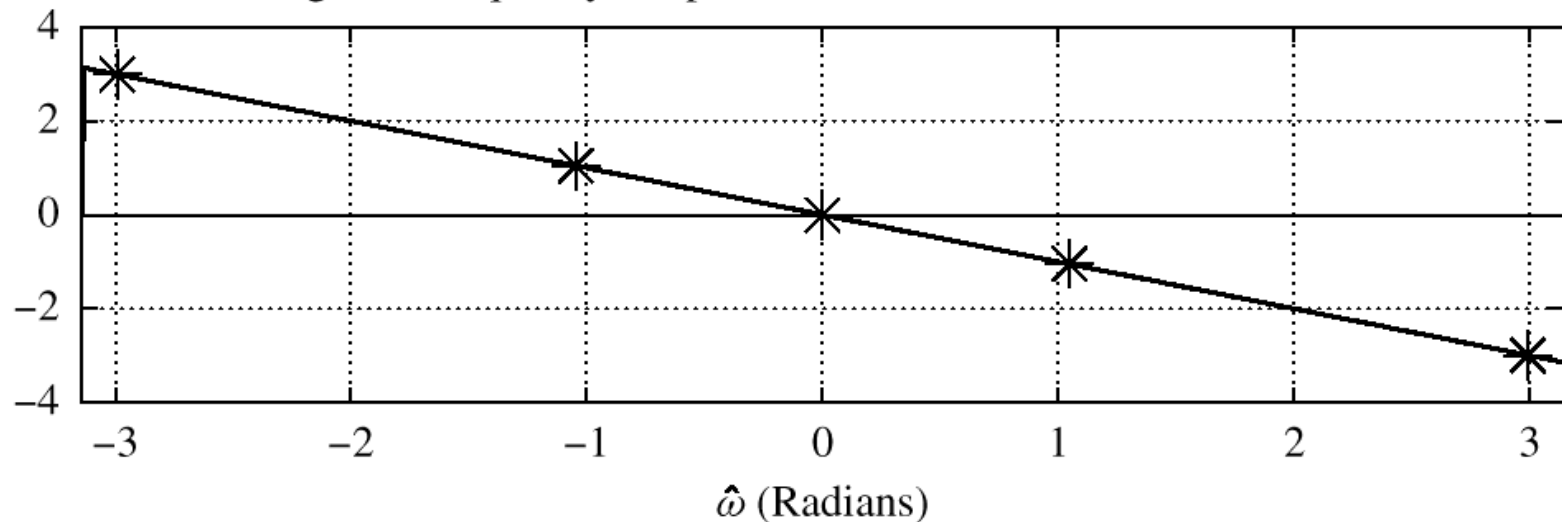
Magnitude of Frequency Response of FIR Filter with Coefficients [1, 2, 1]



$$\angle H(e^{j\hat{\omega}}) = -\hat{\omega}$$

$$\mathcal{H}(\hat{\omega}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$$

Phase Angle of Frequency Response of FIR Filter with Coefficients [1, 2, 1]



Example 2



- Find $y[n]$ for the following inputs when $H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$

1) $x[n] = 2e^{j\pi/4} e^{j(\pi/3)n}$

$$y[n] = (3e^{-j\pi/3}) \times 2e^{j\pi/4} e^{j(\pi/3)n} = 6e^{-j\pi/12} e^{j(\pi/3)n}$$

2) $x[n] = 2 \cos\left(\frac{\pi}{3}n + \frac{\pi}{4}\right) = e^{j(\pi n/3 + \pi/4)} + e^{-j(\pi n/3 + \pi/4)}$

$$y_1[n] = H(e^{j\pi/3}) e^{j(\pi n/3 + \pi/4)} = 3e^{-j(\pi/3)} e^{j(\pi n/3 + \pi/4)}$$

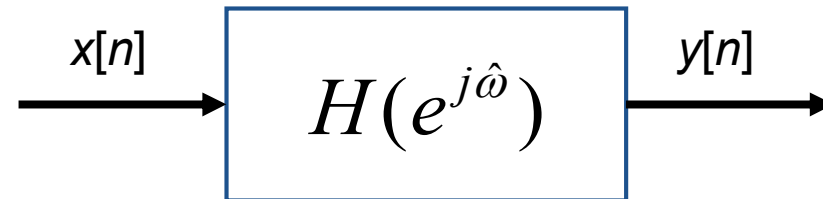
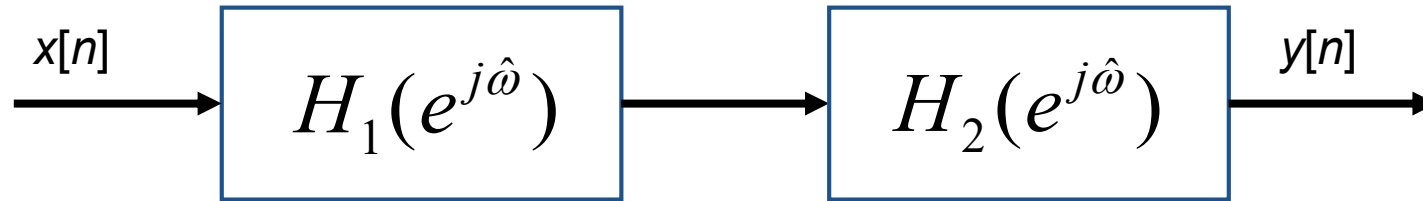
$$y_2[n] = H(e^{-j\pi/3}) e^{-j(\pi n/3 + \pi/4)} = 3e^{j(\pi/3)} e^{-j(\pi n/3 + \pi/4)}$$

$$y[n] = 3e^{j(\pi n/3 - \pi/12)} + 3e^{-j(\pi n/3 - \pi/12)} = 6 \cos\left(\frac{\pi}{3}n - \frac{\pi}{12}\right)$$

Cascaded LTI systems



- Multiply the frequency responses



- Equivalent system

$$H(e^{j\hat{\omega}}) = H_1(e^{j\hat{\omega}})H_2(e^{j\hat{\omega}})$$

L-pt Averager



$$\begin{aligned} H(e^{j\hat{\omega}}) &= \frac{1}{L} \sum_{k=0}^{L-1} e^{-j\hat{\omega}k} \\ &= \frac{1}{L} \left(\frac{1 - e^{-j\hat{\omega}L}}{1 - e^{-j\hat{\omega}}} \right) \\ &= \frac{1}{L} \left(\frac{e^{-j\hat{\omega}L/2} (e^{j\hat{\omega}L/2} - e^{-j\hat{\omega}L/2})}{e^{-j\hat{\omega}/2} (e^{j\hat{\omega}/2} - e^{-j\hat{\omega}/2})} \right) \\ &= \left(\frac{\sin(\hat{\omega}L/2)}{L \sin(\hat{\omega}/2)} \right) e^{-j\hat{\omega}(L-1)/2} = D_L(e^{j\hat{\omega}}) e^{-j\hat{\omega}(L-1)/2} \end{aligned}$$

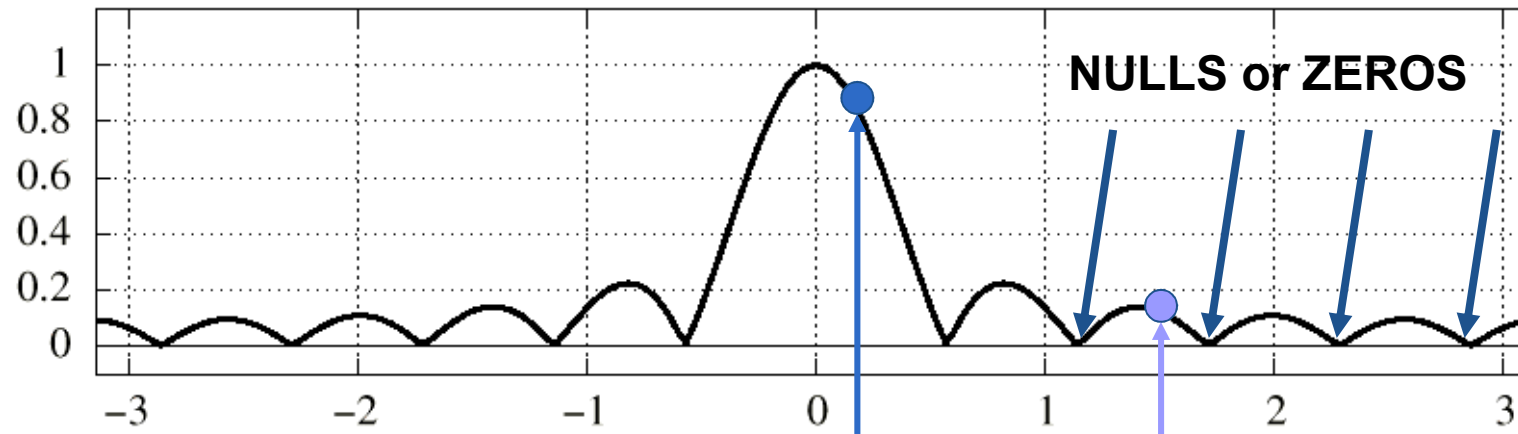
- *Dirichlet* function

$$D_L(e^{j\hat{\omega}}) = \frac{\sin(\hat{\omega}L/2)}{L \sin(\hat{\omega}/2)}$$

11-pt Averager

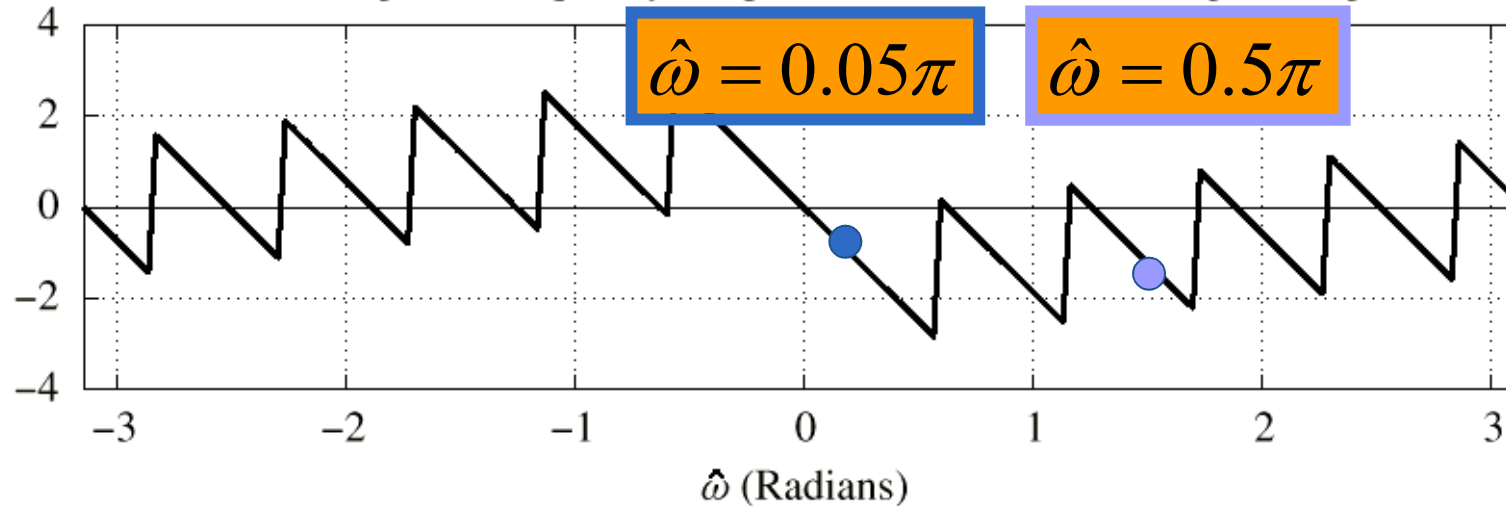


Magnitude of Frequency Response for 11-Point Running Averager



← LPF

Phase Angle of Frequency Response for 11-Point Running Averager



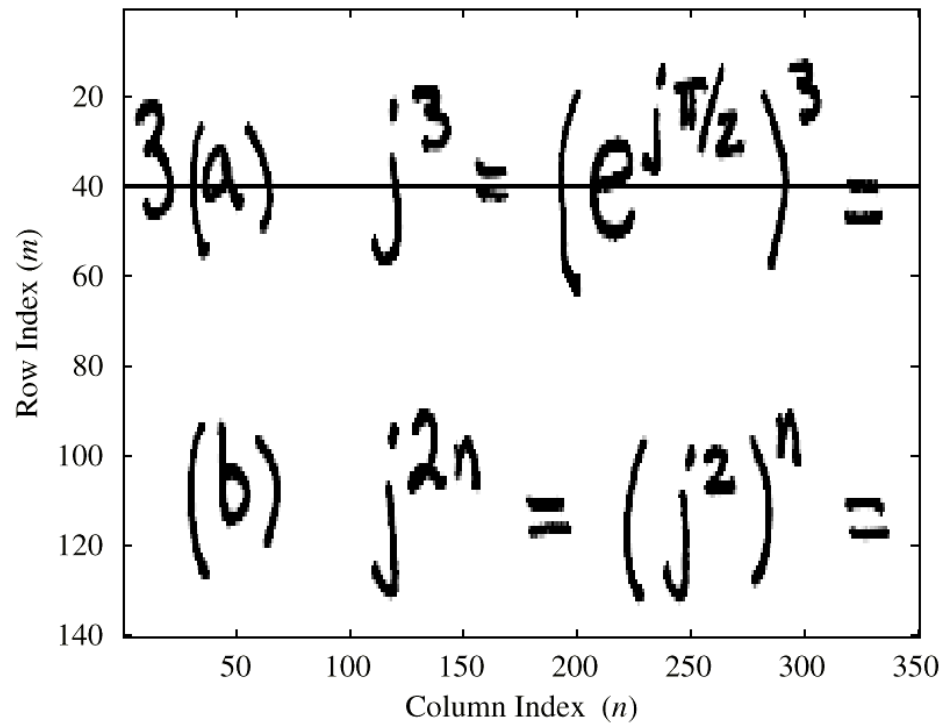
Example: L-pt Averager



- B&W Image

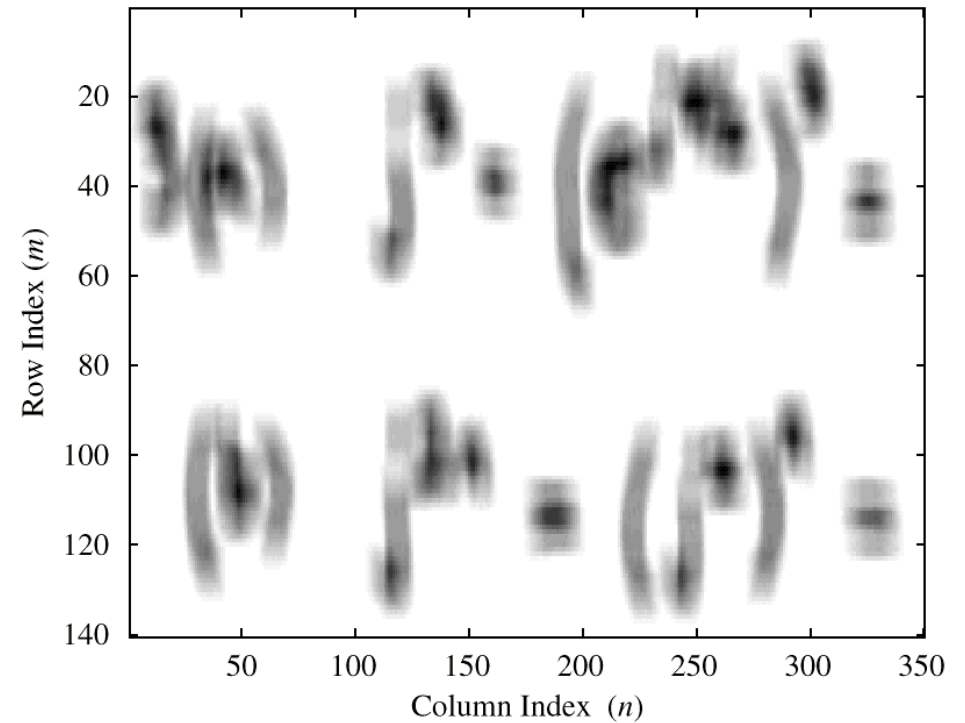
Original

Original Black and White Image



Filtered by 11-pt averager

Row and Column Filtered Image



Summary



- This lecture introduced the concept of FIR filtering.
 - The weighted running average of a finite number of input sequence values defines a discrete-time system.

$$y[n] = \sum_{k=0}^M b_k x[n - k]$$

- The impulse response of an FIR system completely defines the system.
- The frequency-response function is a complete characterization of the behavior of the system for any input that can be represented as a sum of sinusoids.

$$y[n] = H(e^{j\hat{\omega}})x[n]$$